

# GS106 / GS206 SUPERCAPACITOR

## Datasheet Rev 4.3, July 2018

This Datasheet should be read in conjunction with the CAP-XX Supercapacitors Product Guide which contains information common to our product lines.

### Electrical Specifications

The GS106 is a single cell supercapacitor. The GS206 is a dual cell supercapacitor with two GS106 cells in series, so GS206 capacitance = Capacitance of GS106/2 and GS206 ESR = 2 x GS106 ESR.

**Table 1: Absolute Maximum Ratings**

Parameter	Name		Conditions	Min	Typical	Max	Units
Terminal Voltage	V <sub>peak</sub>	GS106		0		2.75	V
		GS206				5.5	
Temperature	T <sub>max</sub>			-40		+70	°C

**Table 2: Electrical Characteristics**

Parameter	Name		Conditions	Min	Typical	Max	Units
Terminal Voltage	V <sub>n</sub>	GS106		0		2.5	V
		GS206		0		5.0	
Capacitance	C	GS106	DC, 23°C	1088	1360	1632	mF
		GS206		544	680	816	
ESR	ESR	GS106	DC, 23°C		20	24	mΩ
		GS206			35	42	
Leakage Current	I <sub>L</sub>		2.3V, 23°C 120hrs		1.5	3	μA
RMS Current	I <sub>RMS</sub>		23°C			7.5	A
Peak Current <sup>1</sup>	I <sub>P</sub>		23°C			30	A

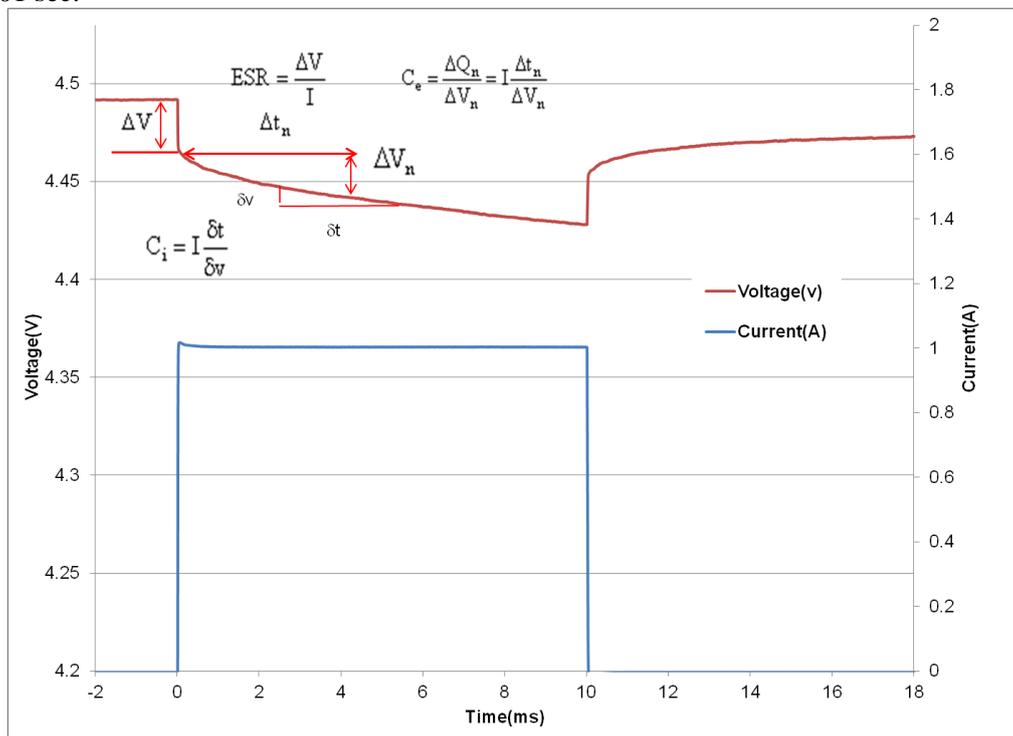
<sup>1</sup>Non-repetitive current, single pulse to discharge fully charged supercapacitor.

**Table 3: Thickness**

GS106F	1.3mm	No adhesive tape on underside of the supercapacitor	GS106G	1.4mm	Adhesive tape on underside, release tape removed
GS206F	2.7mm		GS206G	2.8mm	

## Definition of Terms

In its simplest form, the Equivalent Series Resistance (ESR) of a capacitor is the real part of the complex impedance. In the time domain, it can be found by applying a step discharge current to a charged cell as in Fig. 1. In this figure, the supercapacitor is pre-charged and then discharged with a current pulse,  $I = 1\text{A}$  for duration  $0.01\text{ sec}$ .



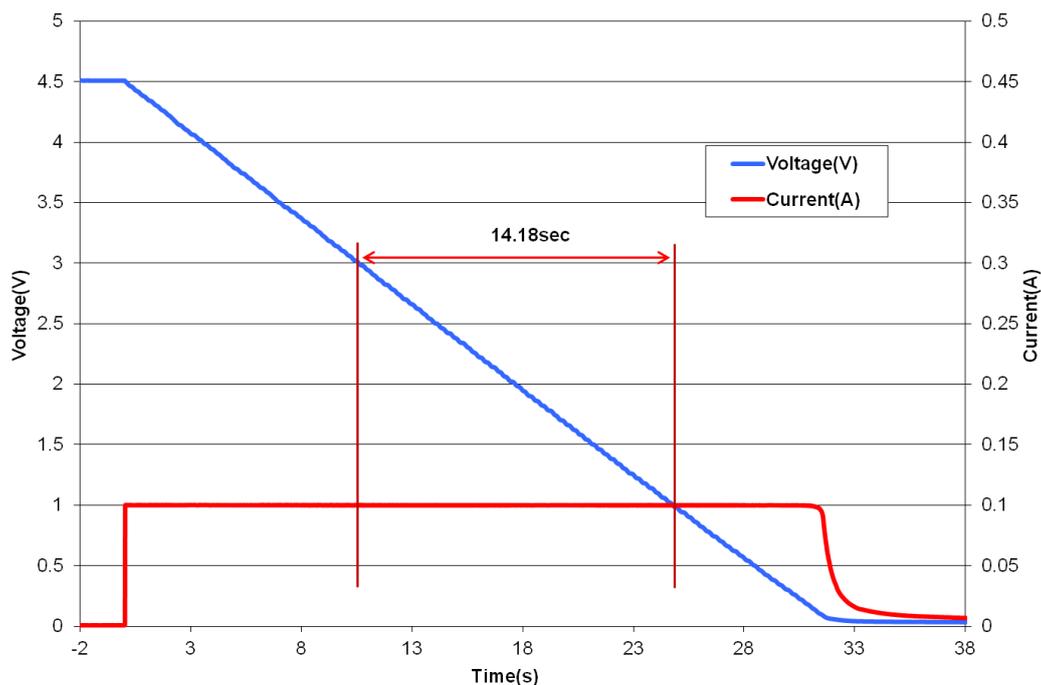
**Figure 1: Effective capacitance, instantaneous capacitance and ESR for a GS206**

The ESR is found by dividing the instantaneous voltage step ( $\Delta V$ ) by  $I$ . In this example  $= (4.49\text{V} - 4.467\text{V})/1\text{A} = 23\text{m}\Omega$ .

The instantaneous capacitance ( $C_i$ ) can be found by taking the inverse of the derivative of the voltage, and multiplying it by  $I$ .

The effective capacitance for a pulse of duration  $\Delta t_n$ ,  $C_e(\Delta t_n)$  is found by dividing the total charge removed from the capacitor ( $\Delta Q_n$ ) by the voltage lost by the capacitor ( $\Delta V_n$ ). For constant current  $C_e(\Delta t_n) = I \times \Delta t_n / \Delta V_n$ .  $C_e$  increases as the pulse width increases and tends to the DC capacitance value as the pulse width becomes very long ( $\sim 10\text{ secs}$ ). After 2msecs, Fig 1 shows the voltage drop  $V_{2\text{ms}} = (4.467\text{ V} - 4.449\text{V}) = 18\text{mV}$ . Therefore  $C_e(2\text{ms}) = 1\text{A} \times 2\text{ms}/18\text{mV} = 111\text{mF}$ . After 10ms, the voltage drop  $= 4.467\text{ V} - 4.428\text{V} = 39\text{mV}$ . Therefore  $C_e(10\text{ms}) = 1\text{ A} \times 10\text{ms}/39\text{mV} = 256\text{mF}$ . The DC capacitance of a GS206 =  $0.68\text{ F}$ . Note that  $\Delta V$ , or  $IR$  drop, is not included because very little charge is removed from the capacitor during this time.  $C_e$  shows the time response of the capacitor and it is useful for predicting circuit behavior in pulsed applications.

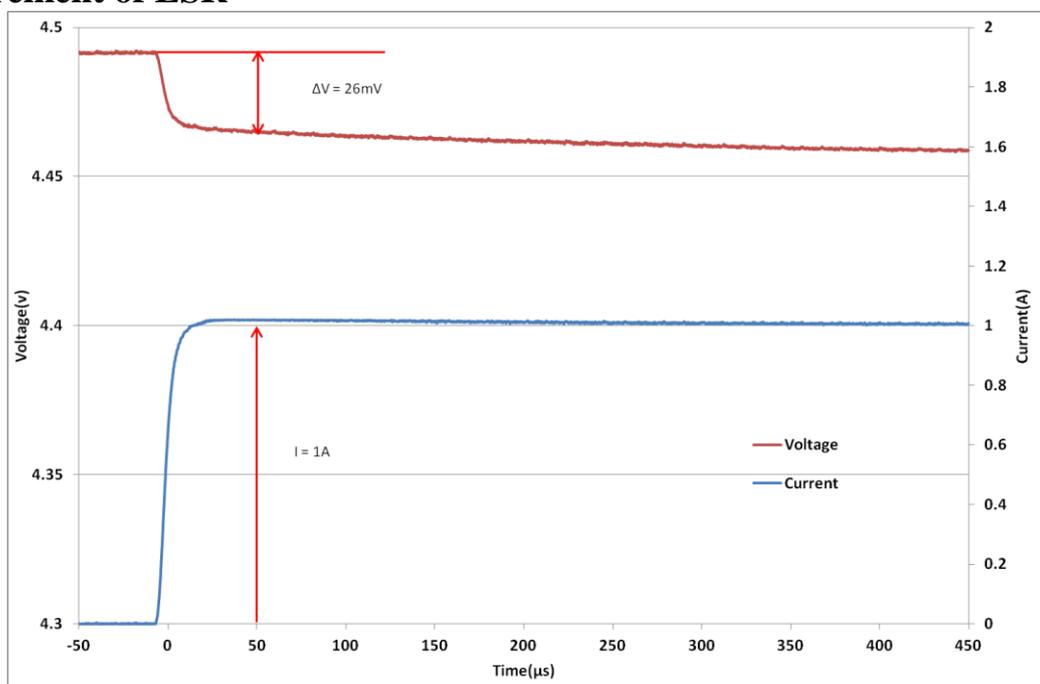
## Measurement of DC Capacitance



**Fig 2: Measurement of DC Capacitance for a GS206**

Fig 2 shows the measurement of DC capacitance by drawing a constant 100mA current from a fully charged supercapacitor and measuring the time taken to discharge from 1.5V to 0.5V for a single cell, or from 3V to 1V for a dual cell supercapacitor. In this case,  $C = 0.1A \times 14.18s / 2V = 709mF$ , which is well within the 680mF +/- 20% tolerance for a GS206 cell.

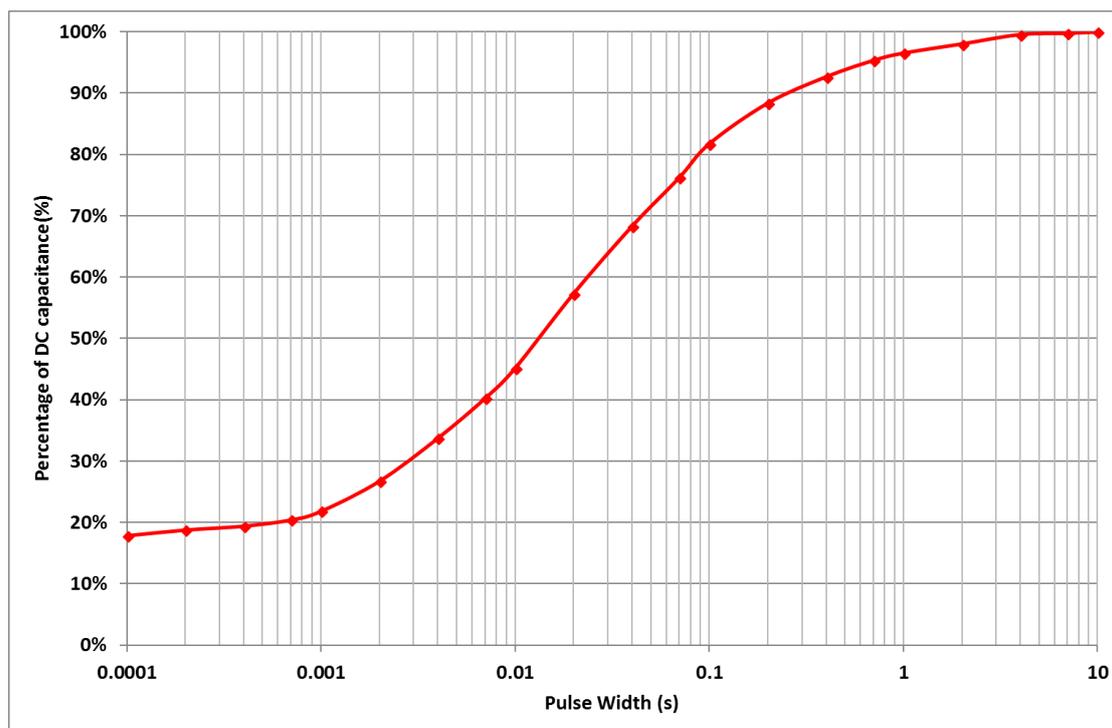
## Measurement of ESR



**Fig 3: Measurement of ESR for a GS206**

Fig 3 shows DC measurement of ESR by applying a step load current to the supercapacitor and measuring the resulting voltage drop. CAP-XX waits for a delay of 50μs after the step current is applied to ensure the voltage and current have settled. In this case the ESR is measured as  $26mV/1A = 26m\Omega$ .

## Effective Capacitance (Ceff)



**Figure 4: Effective Capacitance**

Fig 4 shows the effective capacitance for the GS106, GS206 @ 23°C. This shows that for a 1ms PW, you will measure 22% of DC capacitance or 299mF for a GS106 or 150mF for a GS206. At 10ms you will measure 45% of the DC capacitance, and at 100ms you will measure 82% of DC capacitance. Ceffective is a time domain representation of the supercapacitor's frequency response. If, for example, you were calculating the voltage drop if the supercapacitor was supporting 1A for 10ms, then you would use the  $C_{eff}(10ms) = 45\%$  of DC capacitance = 306mF for a GS206, so  $V_{drop} = 1A \times ESR + 1A \times duration/C = 1A \times 35m\Omega + 1A \times 10ms / 306mF = 68mV$ . The next section on pulse response shows how the effective capacitance is sufficient for even short pulse widths.

## Pulse Response

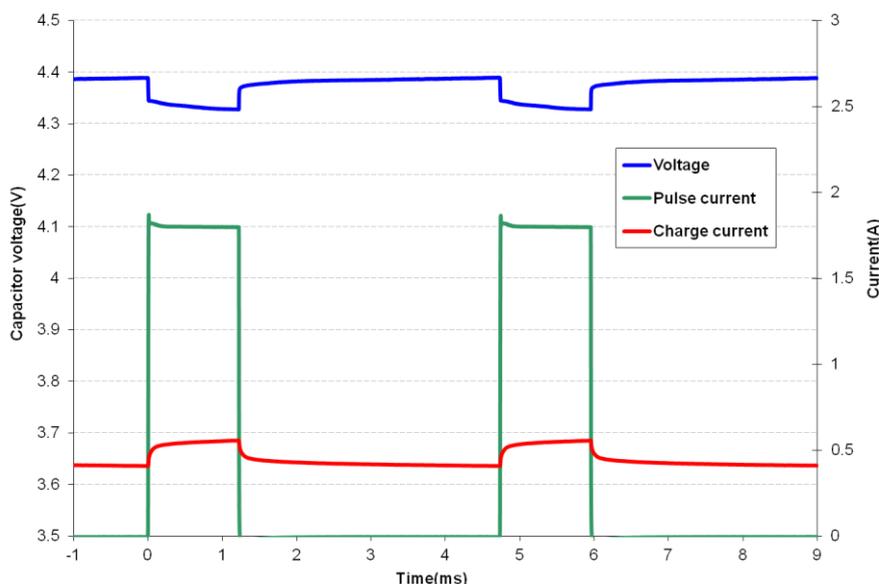


Fig 5 shows that the GS206 supercapacitor does an excellent job supporting a GPRS class 10 pulse train, drawing 1.8A for 1.1ms at 25% duty cycle. The source is current limited to 0.6A and the supercapacitor provides the 1.2A difference to achieve the peak current. At first glance the freq response of Fig 8 indicates the supercapacitor would not support a 1.1ms pulse, but the  $C_{eff}$  of 122mF coupled with the low ESR supports this pulse train with only ~60mV droop in the supply rail.

**Fig 5: GS206 Pulse Response with GPRS Class 10 Pulse Train**

## Accurate Calculation of Voltage Drop for a Pulse Using Ceff

The combination of the method used by CAP-XX to measure ESR and effective capacitance for a given pulsewidth given in Fig 4 enable the accurate calculation of voltage drop for a pulse with current  $I$  and pulsewidth  $T$  as  $V_{drop} = I \cdot [ESR + T/C_{eff}(T)]$ . Using the pulse train of Fig 5 as an example,  $I = 1.8A - 0.6A = 1.2A$ .  $T = 1.1ms$ . Nominal DC capacitance = 680mF, and from Fig 4,  $C_{eff}(1.1ms) = 18\% \times 680mF = 122mF$ . Nominal ESR = 36m $\Omega$ , so  $V_{drop} = 1.2A[0.036\Omega + 0.0011s/0.122F] = 54mV$ . Fig 5 shows a voltage drop = 60mV verifying that the calculation is a good approximation. This avoids the need to run SPICE for a simple calculation.

## DC Capacitance variation with temperature

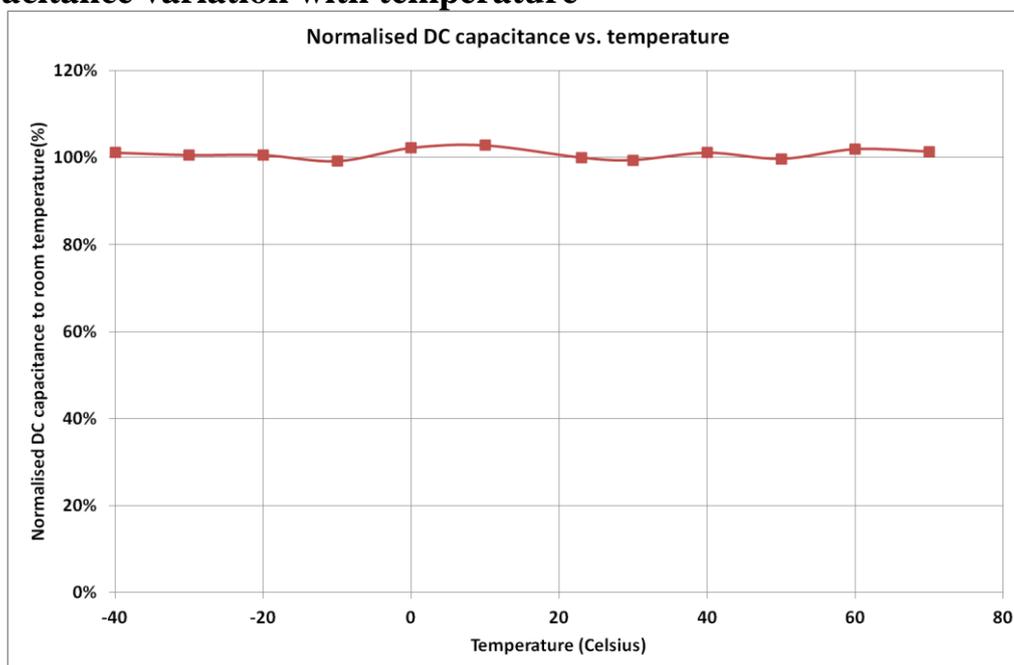


Figure 6: Capacitance change with temperature

Fig 6 shows that DC capacitance is approximately constant with temperature.

## ESR variation with temperature

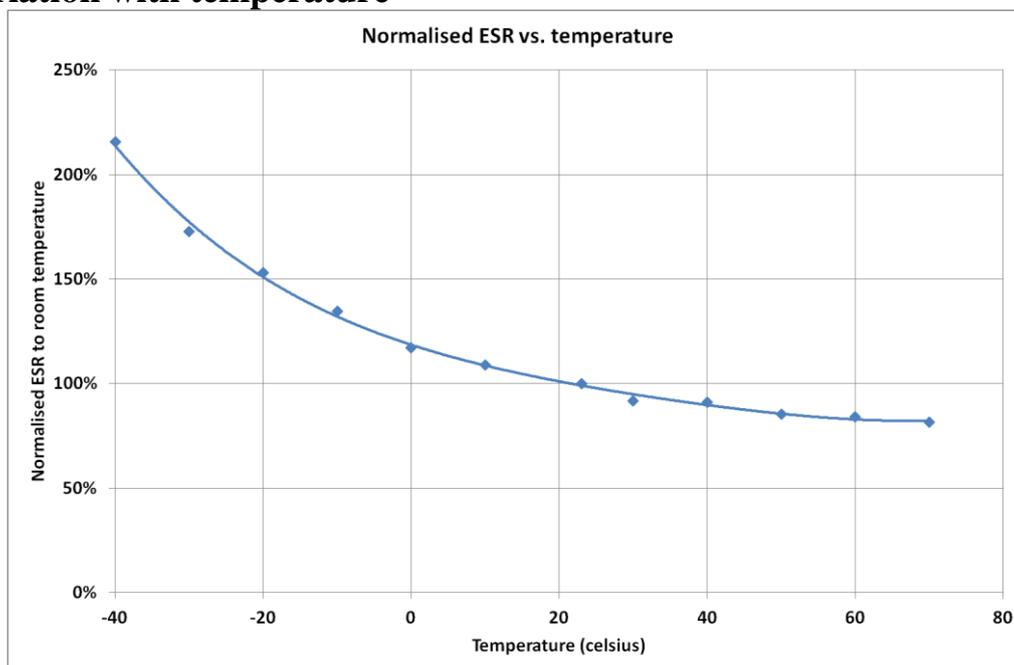


Figure 7: ESR change with temperature

Fig 7 shows that ESR at  $-40^{\circ}\text{C}$  is  $\sim 2.2 \times$  ESR at room temp, and that ESR at  $70^{\circ}\text{C}$  is  $\sim 0.8 \times$  ESR at room temperature.

## Frequency Response

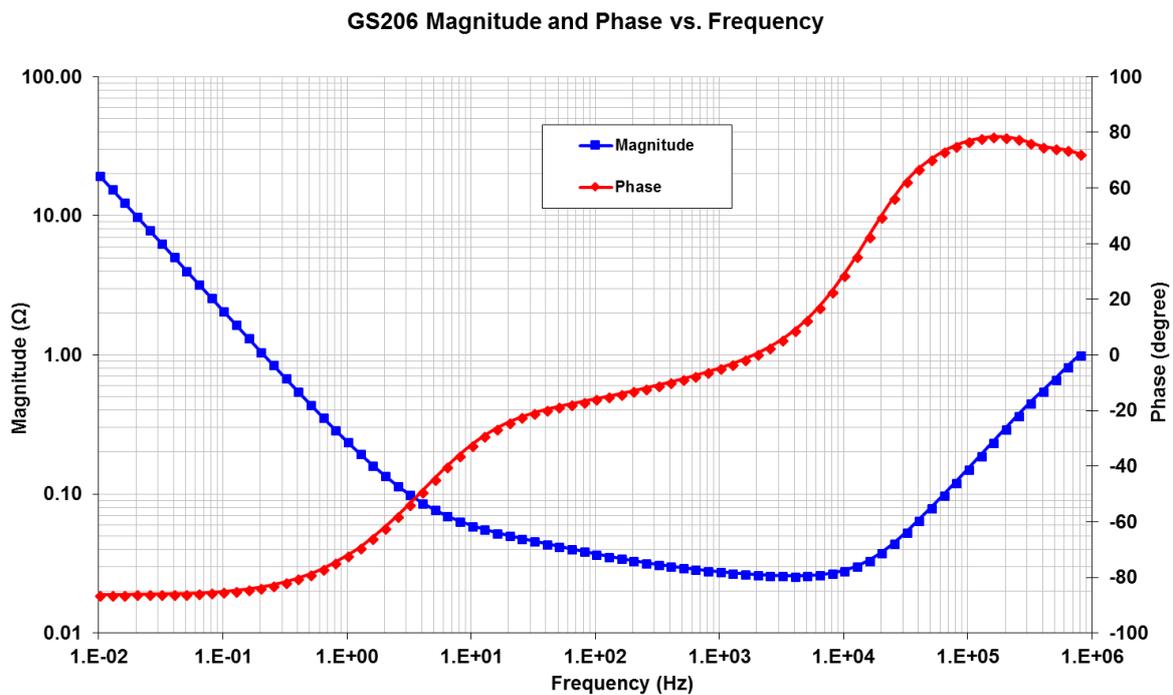


Fig 8: Frequency Response of Impedance (biased at 5V with a 50mV test signal)

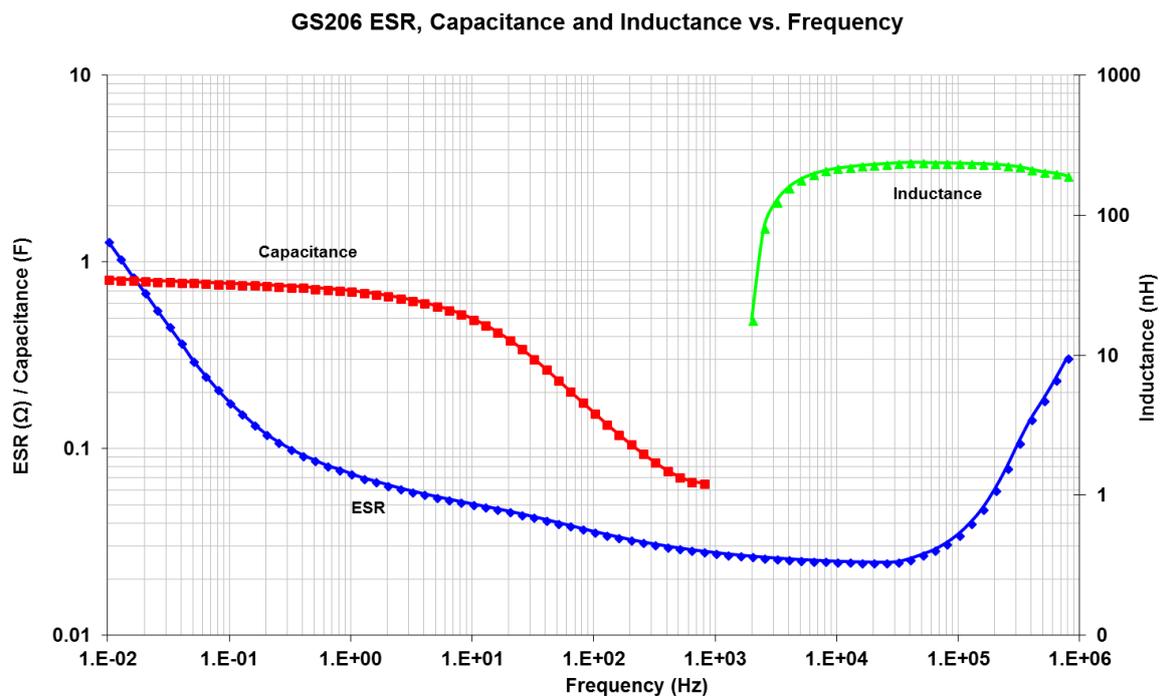
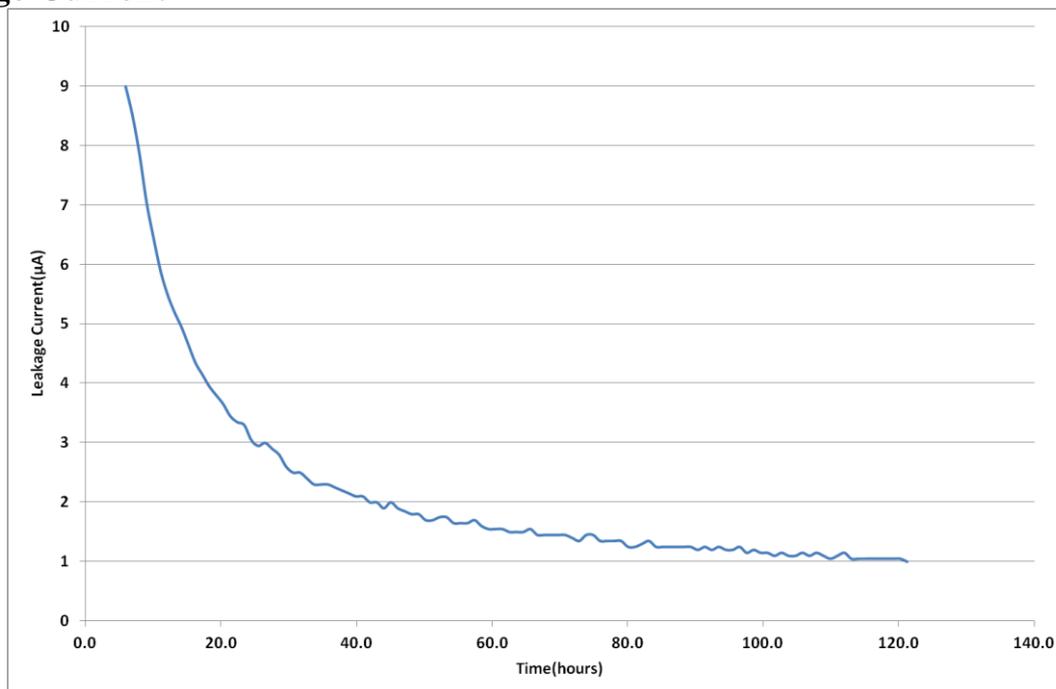


Fig 9: Frequency Response of ESR, Capacitance & Inductance

Fig 8 shows the supercapacitor behaves as an ideal capacitor until approx 5 Hz when the magnitude no longer rolls off proportionally to  $1/\text{freq}$  and the phase crosses  $-45^{\circ}$ . Performance of supercapacitors with frequency is complex and the best predictor of performance is Fig 4 showing effective capacitance as a function of pulsewidth.

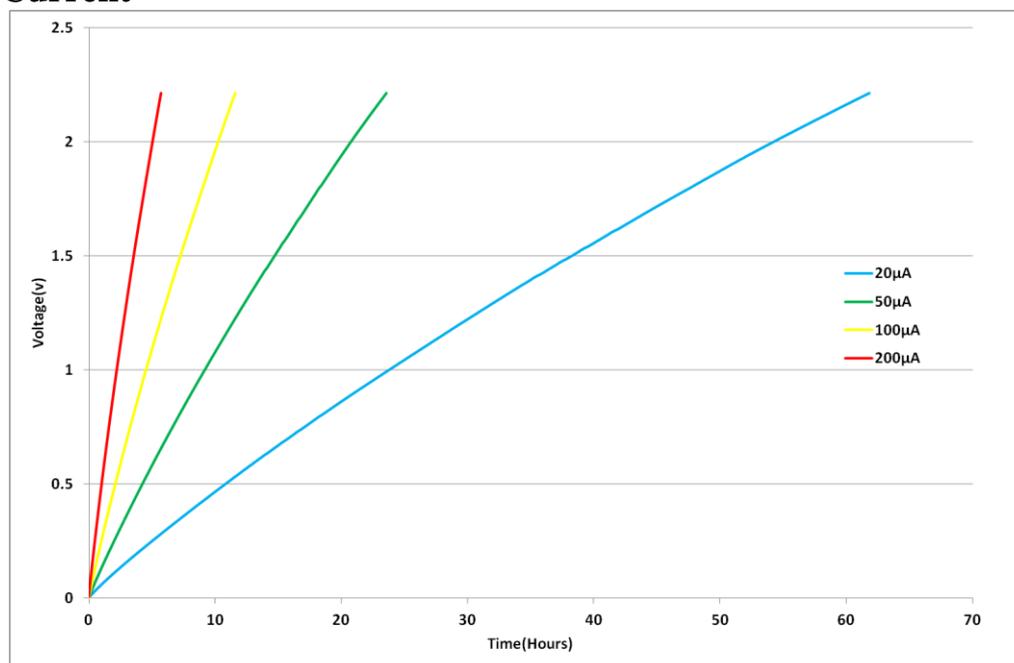
## Leakage Current



**Fig 10: Leakage Current**

Fig 10 shows the leakage current for GS106 at room temperature. The leakage current decays over time, and the equilibrium value leakage current will be reached after ~120hrs at room temperature. The typical equilibrium leakage current is 1µA at room temperature. At 70°C leakage current will be ~10µA.

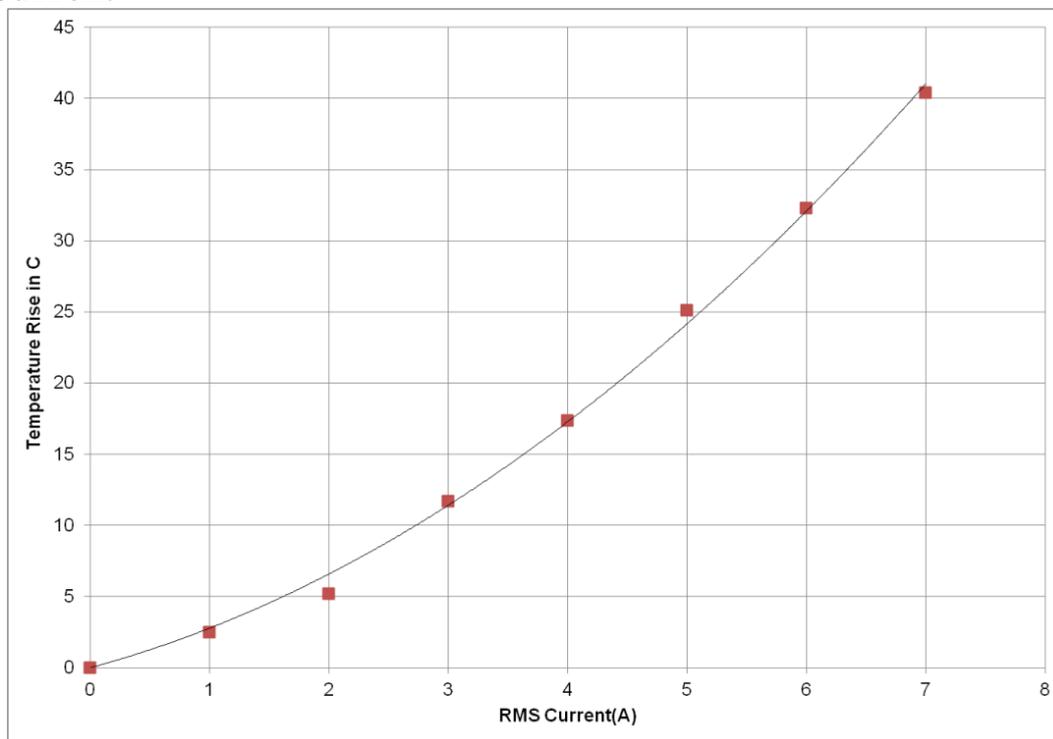
## Charge Current



**Fig 11: Charging a GS106 with low current**

The corollary to the slow decay in leakage currents shown in Fig 10 is that charging a supercapacitor at very low currents takes longer than theory predicts. At higher charge currents, the charge rate is as theory predicts. For example, it should take  $1.36 \text{ F} \times 2.2\text{V} / 0.00002\text{A} = 41.6\text{hrs}$  to charge a 1.36F supercapacitor to 2.2V at 20µA, but Fig 11 shows it took 64hrs. At 100µA charging occurs at a rate close to the theoretical rate.

## RMS Current



**Fig 12: Temperature rise in GS206 with RMS current**

Continuous current flow into/out of the supercap will cause self heating, which limits the maximum continuous current the supercapacitor can handle. This is measured by a current square wave with 50% duty cycle, charging the supercapacitor to rated voltage at a constant current, then discharging the supercapacitor to half rated voltage at the same constant current value. For a square wave with 50% duty cycle, the RMS current is the same as the current amplitude. Fig 12 shows the increase in temperature as a function of RMS current. From this, the maximum RMS current in an application can be calculated, for example, if the ambient temperature is 40°C, and the maximum desired temperature for the supercapacitor is 70°C, then the maximum RMS current should be limited to 5.5A, which causes a 30°C temperature increase.

## CAP-XX Supercapacitors Product Guide

Refer to the package drawings in the CAP-XX Supercapacitors Product Guide for detailed information of the product's dimensions, PCB landing placements, active areas and electrical connections.

Refer to the CAP-XX Supercapacitors Product Guide for information on endurance and shelf life, transportation and storage, assembly and soldering, safety and RoHS/EREACH certification.